

LOW VOLTAGE RIDE THROUGH OF A GRID CONNECTED DOUBLY FED INDUCTION GENERATOR WITH FIELD ORIENTED CONTROL

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ABSTRACT

This paper deals with a new solution for Low voltage ride through of a grid connected DFIG with field-oriented control (vector control). Control itself acting a very vital part in drives and wind turbine technology. Control of the DFIG when generating energy in a wind turbine is essential and obvious. The control techniques used for DFIG are vector control (field-oriented control) and direct control technique. In vector control of a DFIG, the components of the d and the q axis components of the rotor currents and voltages are measured with PI controllers. If a reference frame oriented with the stator flux is used, the active and reactive power flows of the stator can be measured by independently by resources of quadrature and the direct current, respectively. By means of the direct current, we can control the active power whereas the reactive power can be measured via the quadrature current. A Dynamic modelling of DFIG is presented. In our power system, in way to keep constant power, voltage and frequency we use DFIG. In this a complete study is done to study the field-oriented control characteristics of DFIG by using rotor current control loops and grid voltage-oriented vector control. This system modelled in MATLAB/Simulink environment.

KEYWORDS: Low voltage ride through, Doubly-fed induction generator (DFIG), Field oriented control, Wind turbine power system.

Article History

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INTRODUCTION FOR DFIG WIND TURBINE

The DFIG is a variable speed wind turbine together with induction machine where also rotor is linked to grid. Power is also provided from grid (or) distributed to grid through rotor. This power is called slip power. Frequency of slip power is varied in such a way that the rotor field frequency is kept constant. This difference of frequency of slip power is recognized by two power electronics back-back converters. Bidirectional flow of power in back-back converters gives work in sub synchronous mode as well as over synchronous mode. In DFIG the back-back converter is containing of one machine-side converter, a DC link capacitor and grid side converter. The part of machine-side converter is to switch speed (or) torque of DFIG and machine power factor. The role of grid side converter is to minimise DC link capacitor voltage ripple and also injects reactive power to the line. Operating speed range doesn't exceed 40% of synchronous speed. The cost and losses of converters in DFIM are lower in comparison to full power converters. DFIG can activate close to unity power factor. The

below figure shows the block diagram of a Variable speed wind turbine system with DFIG.

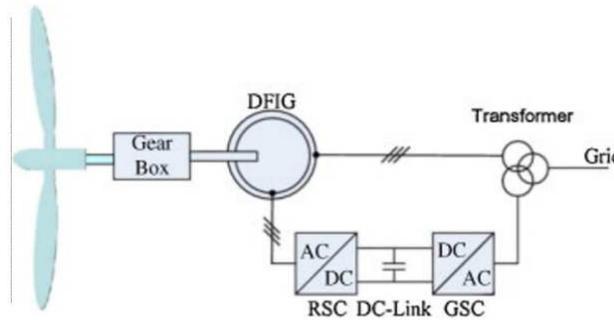


Figure 1: Variable Speed Wind Turbine with DFIM.

DYNAMIC MODELLING OF DFIG

This system consists of a wind turbine with DFIG. The principle of DFIG is that rotor windings are linked to the grid via slip rings and back-back voltage source converters that controls both rotor and grid currents. Thus, rotor frequency can freely differ from the grid frequency. By using converter to control rotor currents, it is possible to adjust active and reactive power fed to grid from stator independently if generator turning speed.

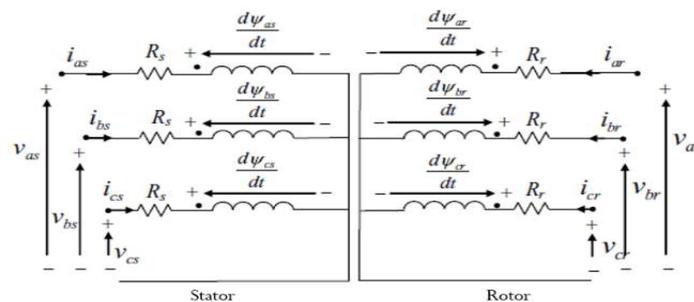


Fig:Electric Equivalent circuit of DFIG

Figure 2

Applying KVL to Stator side

$$v_{as}(t) = R_s i_{as}(t) + \frac{d\psi_{as}(t)}{dt}$$

$$v_{bs}(t) = R_s i_{bs}(t) + \frac{d\psi_{bs}(t)}{dt}$$

$$v_{cs}(t) = R_s i_{cs}(t) + \frac{d\psi_{cs}(t)}{dt}$$

Similarly, to Rotor side

$$v_{ar}(t) = R_r i_{ar}(t) + \frac{d\psi_{ar}(t)}{dt}$$

$$v_{br}(t) = R_r i_{br}(t) + \frac{d\psi_{br}(t)}{dt}$$

$$v_{cr}(t) = R_r i_{cr}(t) + \frac{d\psi_{cr}(t)}{dt}$$

αβ Model

In developing the dynamic αβ model of the DFIG, space vector theory is useful to the basic electric equations of machine. The stator reference frame (α-β) is a stationary reference frame, the rotor reference frame (DQ) rotates at ωm and the synchronous reference frame (dq) rotates at ωs. Subscripts “s”, “r”, “a” are used to denote that one space vector is reference to stator, rotor and synchronous reference frame respectively. By using direct and inverse rotational transformation, a space vector can be represented in any of these frames.

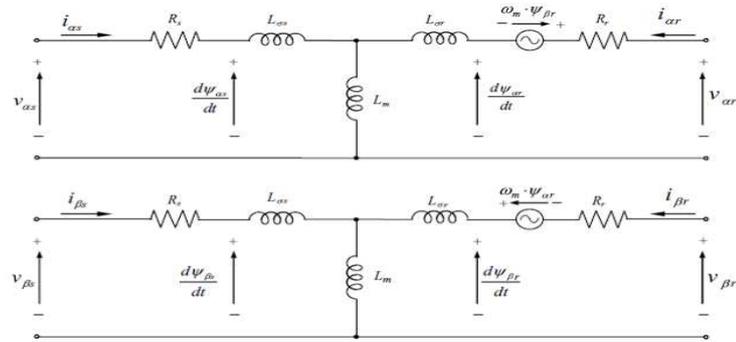


Figure 3

Applying KVL to Stator side and Rotor side using space vector notation

$$\vec{v}_s^s = R_s \vec{i}_s^s + \frac{d\vec{\psi}_s^s}{dt}$$

$$\vec{v}_r^r = R_r \vec{i}_r^r + \frac{d\vec{\psi}_r^r}{dt}$$

The flux is given by

$$\Psi = L * i$$

In space vector notation is given by

$$\vec{\psi}_s^s = L_s \vec{i}_s^s + L_m \vec{i}_r^r \quad \text{Stator reference frame}$$

$$\vec{\psi}_r^r = L_m \vec{i}_s^s + L_r \vec{i}_r^r \quad \text{Rotor reference frame}$$

Power expressions

$$S = P + JQ$$

By using space vector notation, the apparent power is given by

$$S = 3/2 (V * i)$$

$$S = 3/2[(V_\alpha * i_\alpha + i_\beta * V_\beta) + J (V_\beta * i_\alpha - V_\alpha * i_\beta)]$$

Active power is

$$P = 3/2(V \alpha * i \alpha + i \beta * V \beta)$$

Reactive power is

$$Q = 3/2(V \beta * i \alpha - V \alpha * i \beta)$$

The electric powers on stator and rotor side are calculated as follows

$$P_s = 3/2(V \alpha_s * i \alpha_s + V \beta_s * i \beta_s)$$

$$P_r = 3/2(V \alpha_r * i \alpha_r + V \beta_r * i \beta_r)$$

$$Q_s = 3/2(V \beta_s * i \alpha_s - V \alpha_s * i \beta_s)$$

$$Q_r = 3/2(V \beta_r * i \alpha_r - V \alpha_r * i \beta_r)$$

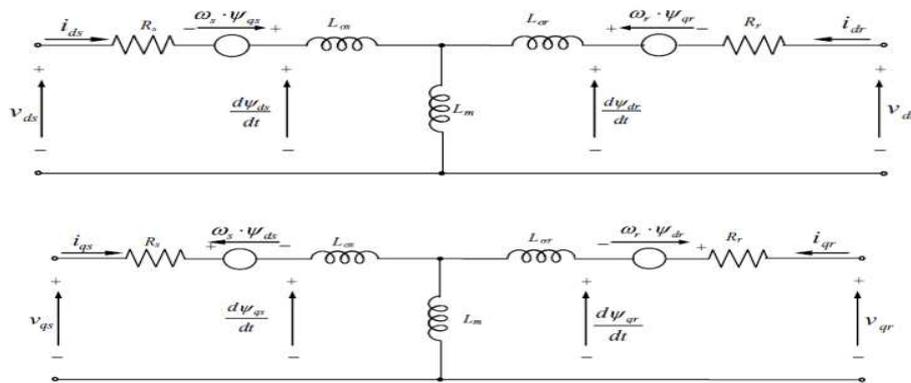
Electromagnetic torque can be found

$$T_{em} = 3/2 * P * (i_m) * [(\Psi_r \cdot i_r^*)]$$

$$T_{em} = 3/2 * P * (\Psi \beta_r * i \alpha_r - \Psi \alpha_r * i \beta_r)$$

DQ Model

The space vector model of the DFIG can also be represented in a synchronously rotating frame.



This model of DFIG is derived using space vector notation in synchronous reference frame

Figure 4

The stator and rotor voltages is given by

$$\vec{v}_s^a = R_s \vec{i}_s^a + \frac{d\vec{\psi}_s^a}{dt} + j\omega_s \vec{\psi}_s^a$$

$$\vec{v}_r^a = R_r \vec{i}_r^a + \frac{d\vec{\psi}_r^a}{dt} + j(\omega_s - \omega_m) \vec{\psi}_r^a$$

Flux expression is given by

$$\vec{\psi}_s^a = L_s \vec{i}_s^a + L_m \vec{i}_r^a$$

$$\vec{\psi}_r^a = L_m \vec{i}_s^a + L_r \vec{i}_r^a$$

The power expression in DQ model is given by

$$P_s = 3/2(V_{ds} * i_{ds} + V_{qs} * i_{qs})$$

$$P_r = 3/2(V_{dr} * i_{dr} + V_{qr} * i_{qr})$$

$$Q_s = 3/2(V_{qs} * i_{ds} - V_{ds} * i_{qs})$$

$$Q_r = 3/2(V_{qr} * i_{dr} - V_{dr} * i_{qr})$$

Torque expression is given by

$$T_{em} = 3/2 * P(L_m/L_s) * [(\Psi_{qs} * i_{dr} - \Psi_{ds} * i_{qr})]$$

FIELD-ORIENTED CONTROLLING OF DFIG

Rotor Current Control Loops

Among the different control methods that have been developed for the DFIG, only the field-oriented control method is studied in this section. The FOC of the DFIG is performed in a synchronously rotating dq frame, in which the d-axis is aligned, in this case, with the stator flux space vector. It will be shown later that the direct rotor current is proportional to the stator reactive power and that the quadrature rotor current is proportional to the torque or active stator power.

$$V_{dr} = R_r * i_{dr} + \sigma L_r \frac{d}{dt} i_{dr} - \omega_r \sigma L_r * i_{qr} + L_m/L_s \frac{d}{dt} \psi_s^a$$

$$V_{qr} = R_r * i_{qr} + \sigma L_r \frac{d}{dt} i_{qr} + \omega_r \sigma L_r * i_{dr} + \omega_r L_m/L_s \psi_s^a$$

Assuming that the voltage drop in the stator resistance is small, the stator flux is constant because the stator is connected directly to the grid at constant AC voltage. It is possible to perform dq rotor currents control, simply by using a regulator for each current component. For the reference frame transformation, the angle θ_r must be estimated. The control must be performed in dq coordinates, but then the rotor voltage and currents must be transformed into DQ coordinates. A simple phase-locked loop (PLL) can be used to perform the stator voltage grid synchronization, providing robustness to the estimation and a rejection of small disturbances or harmonics. The current loops work with the rotor currents referred to the stator side, while the conversion to rotor-referred quantities is performed at the measurement stage for the currents and before the creation of the pulses for the converter for the voltages.

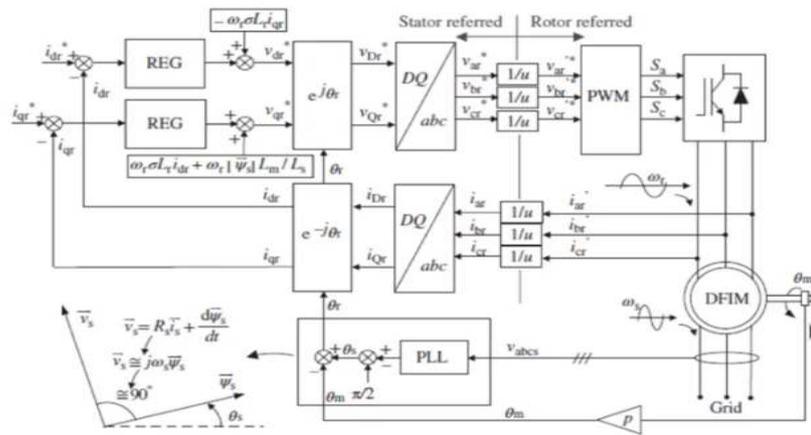


Figure 5: Current Control Loops of DFIM.

Power and Speed Control Loops

Once the current control loops and the flux angle calculation have been studied, the complete control system can be introduced. As the d-axis of the reference frame is aligned with the stator flux space vector, the torque expression in the dq frame can be simplified as follows:

$$T_{em} = 3/2 p L_m L_s (\psi_{qs} i_{dr} - \psi_{ds} i_{qr}).$$

$$T_{em} = 3/2 p L_m L_s |\psi^{***} s| i_{qr} \Rightarrow T_{em} = K_t i_{qr}$$

This means that the q rotor current component is proportional to the torque, that is, with i_{qr} it is possible to control the torque and, consequently, the speed of the machine if the application requires it. In a similar way, by developing the stator reactive power expression in the dq frame, we obtain a compact expression, which reveals that i_{dr} is responsible of Q_s.

$$Q_s = 3/2 (v_{qs} i_{ds} - v_{ds} i_{qs}).$$

$$Q_s = - 3/2 L_m L_s |\psi^{***} s| [i_{dr} - |\psi_s| L_m] \Rightarrow Q_s = K_Q [i_{dr} - |\psi_s| L_m]$$

Therefore, it can be seen that both rotor current components independently allow us to control the torque and reactive stator power. However, with the Q_s loop, it is possible to control the magnetizing of the machine. As discussed before, because the stator of the machine is connected directly to the grid, the stator flux amplitude is constant and provided by the grid voltage: $|\psi^{***} s| = \psi_{ds} = L_s i_{qs} + L_m i_{dr} \psi_{ds} = 0 = L_s i_{qs} + L_m i_{qr}$. The stator flux level $|\psi_s|$ must be created by choosing i_{ds} and i_{dr}, distributing thus the required amount of current between the rotor and the stator. The equivalent closed-loop systems of the Q_s and ω_m loops, assuming that the current loops have been tuned much faster than the external loops and neglecting converter dynamics or measurement and computing delays. It can be seen that the simplified closed-loop systems yield into a first- and second order system that can be tuned by choosing the appropriate gains of the PI regulators. Finally, the most representative magnitudes of a vector controlled DFIM, operating at constant torque in motor mode at variable speed. The stator voltage is kept constant owing to the direct grid connection, while the stator currents are also constant because T_{em} and Q_s are maintained constant. The speed ramp performed at the middle of the experiment provokes the variation of the rotor voltage and currents, which yields a variation of the rotor active and reactive powers.

FOC OF THE GRID SIDE SYSTEM

Control is a necessary part of the grid side system. In this section, a FOC based schema is studied. This control method is widely extended among the control strategies for grid connected converters. It provides good performance characteristics with reasonably simple implementation requirements. The field-oriented control method follows the philosophy of representing the system that is going to be controlled in our case the grid side system in a space vector form.

Grid Voltage Oriented Vector Control:

The grid side converter is in charge of controlling part of the power flow of the DFIG. The power generated by the wind turbine is partially delivered through the rotor of the DFIG. This power flow that goes link and finally is transmitted by the grid side converter to the grid. The simplified block diagram of the grid side system, together with a schematic of its control block diagram, is given in below.

The pulses for the controlled switches (S_{a_g} , S_{b_g} , S_{c_g}) of the 2L-VSC, that is, the output voltage of the converter, are generated in order to control the DC bus voltage (V_{bus}) of the DC link and the reactive power exchanged with the grid (Q_g). This is done, in general, according to a closed loop control law. Some typical controls are vector control or direct power control. However, this section only studies the grid voltage-oriented vector control (GVOVC). Control of V_{bus} is necessary since, the DC link is mainly formed by a capacitor. Thus, the active power flow through the rotor must cross the DC link and then it must be transmitted to the grid. Therefore, by only controlling the V_{bus} variable to a constant value, this active power flow through the converters is ensured, together with a guarantee that both grid and rotor side converters have available the required DC voltage to work properly. Therefore, the grid voltage-oriented vector control (GVOVC) block diagram is shown in below.

Figure 7 shows the V_{bus} and Q_g references, it creates pulses for the controlled switches S_{a_g} , S_{b_g} , and S_{c_g} . Thus, the modulator creates the pulses S_{a_g} , S_{b_g} , S_{c_g} from the abc voltage references for the grid side converter: V^*_{af} , V^*_{bf} , and V^*_{cf} . In this way, these abc voltage references are first created in dq coordinates (v^*_{df} , v^*_{af}), then transformed to ab coordinates (v^*_{af} , v^*_{bf}), and finally generate the abc voltage references. Then, the dq voltage references (v^*_{df} , v^*_{af}) are independently created by the dq current (i^*_{dg} , i^*_{qg}) controllers.

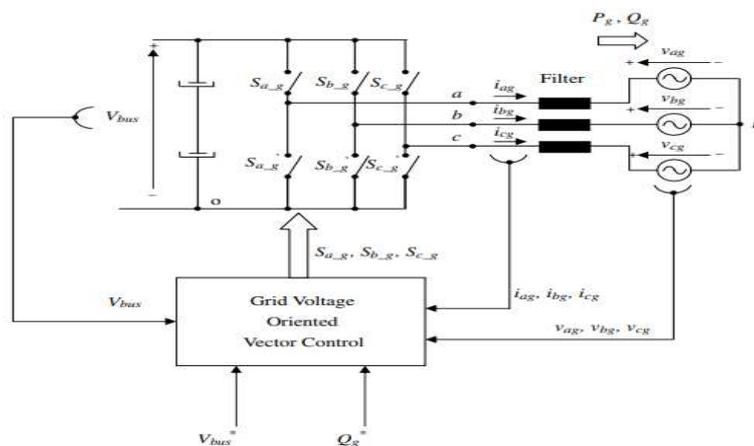


Figure 6: Grid Side System Control.

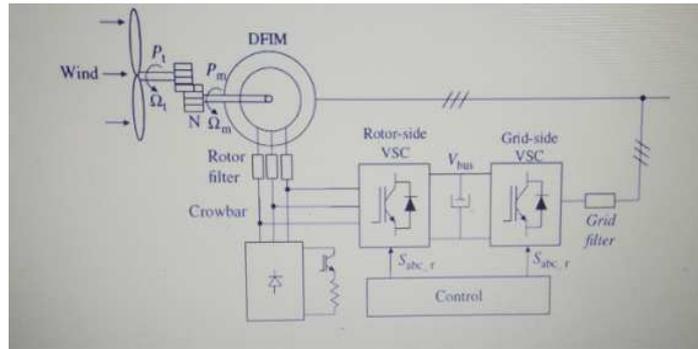


Figure 8

Asymmetrical Voltage Dips

A voltage dip is not same in all the three phases such a dip is said to be unbalanced or asymmetrical dip. The most common faults for electrical networks are single line to ground short circuit, short circuit between two lines.

Asymmetrical Voltage Dips in Wind Turbine Based DFIG

In this the control strategy used is dual vector control. We need positive and negative sequence of voltage and currents to the DFIG we should include negative sequence current loops in order to proper control of the generator. The below figure shows the dual vector control block diagram of DFIG.

An unbalanced voltage implies the presence of a negative sequence that causes oscillations in the torque, over currents in the stator and over voltages in the rotor. The most of these problems can be overcome by negative sequence in the current references. The current references are the addition of two sequences- one synchronized with the positive sequence of the grid voltage, and the other synchronized with the negative one. The original control loop is then substituted by two control loops, one working in a positive rotating reference frame and the other working in an inversely rotating frame. This method is called dual control method. The below figure shows a typical schema of a dual control. The boxes in light gray are the two current controls: the top one regulates the positive sequence, while the bottom one regulates the negative sequence. The measured current must be split into its two sequences before introducing it to the current regulators. Represent the voltage dip a symmetrical voltage dip that one to be emulate in our system. We left positive sequence we use fundamental and harmonic generation. In order to create a symmetry, we are going to use first order with an amplitude of 0.2 for resistance and phase shift 0 for negative sequence voltage. For negative sequence we represent the control loops at grid voltage.

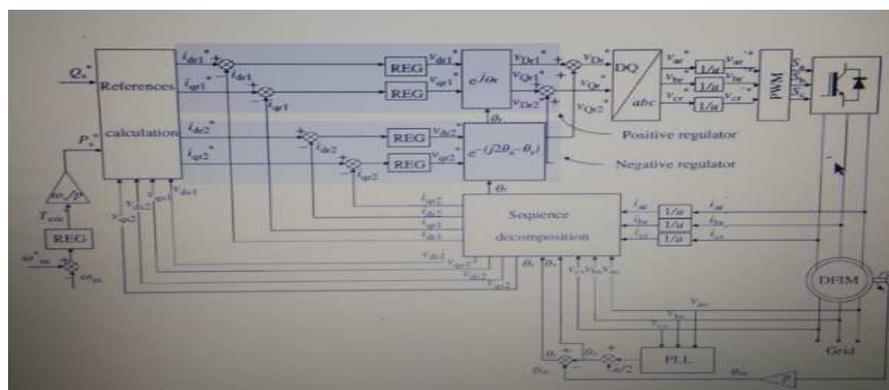


Figure 9: Dual Vector Control Block Diagram of DFIG.

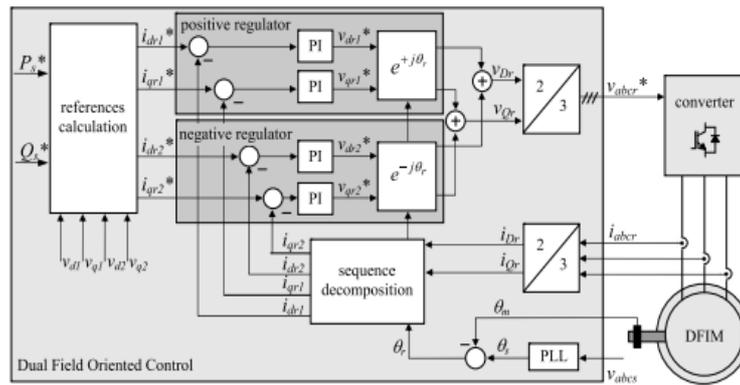


Figure 10: Schematic Diagram of Dual Control.

SIMULINK DIAGRAM

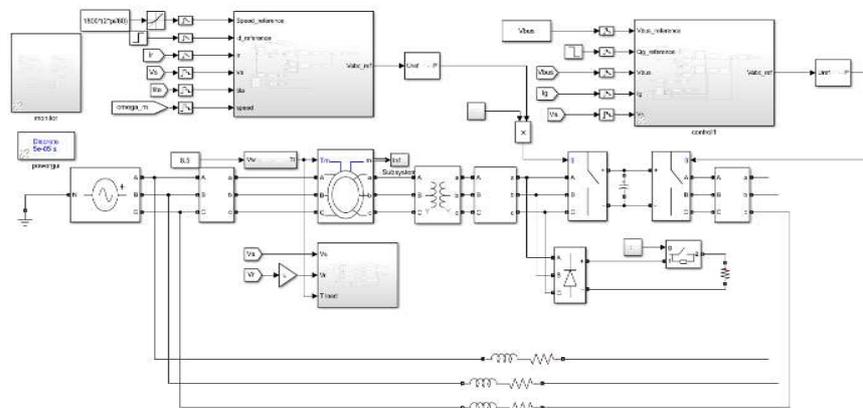


Figure 11

SIMULATION RESULTS

Consider a DFIG based wind turbine system having Normal power of 2MW, Terminal voltage of 690V is connected to 120Kv,50Hz Grid via Two step up Transformer of rating 690V/25Kv and 25Kv/120Kv. The normal wind speed is 10m/sec. No of poles pairs is 2. Nominal DC link voltage is 1.2Kv. DC link capacitance is 10mF. Stator resistance is 0.0048 p.u and Rotor resistance is 0.0044 p.u. Mutual inductance is 6.77 p.u inertia coefficient is H=5sec.

OUTPUT WAVEFORMS

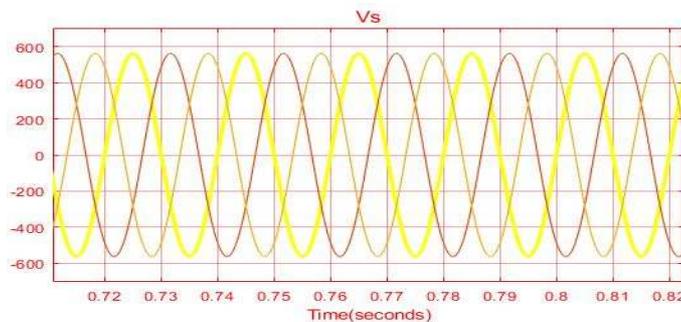


Figure 12

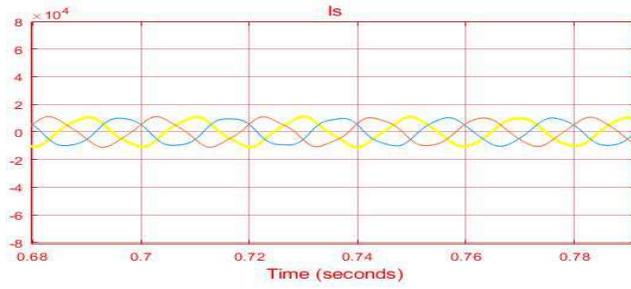


Figure 13

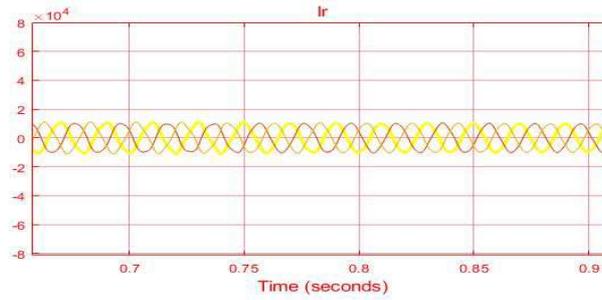


Figure 14

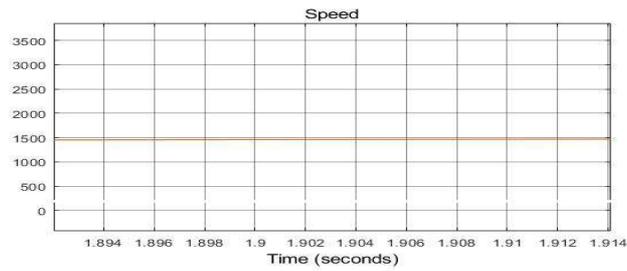


Figure 15

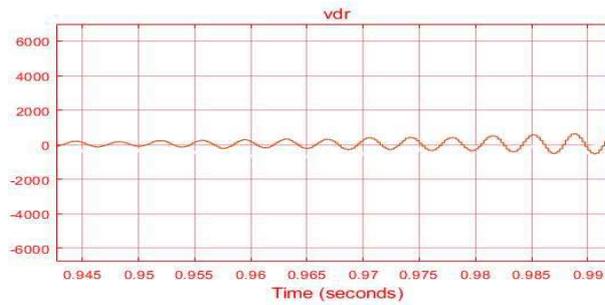


Figure 16

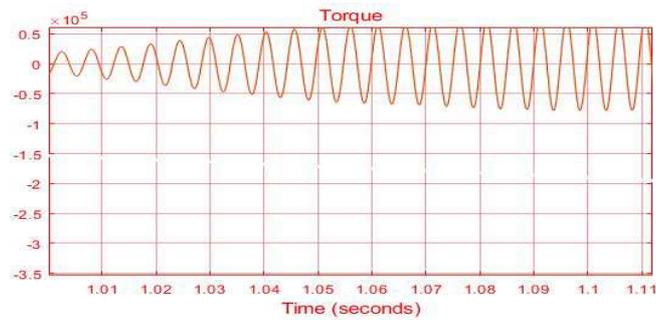


Figure 17

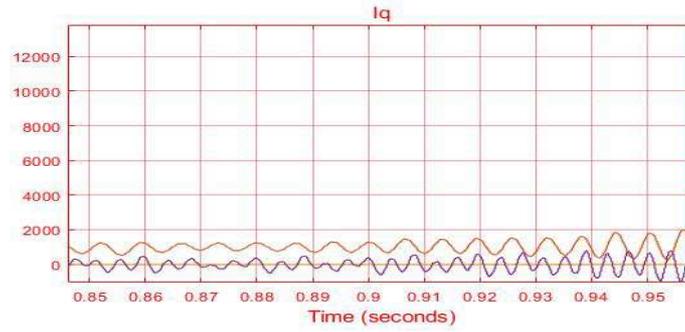


Figure 18

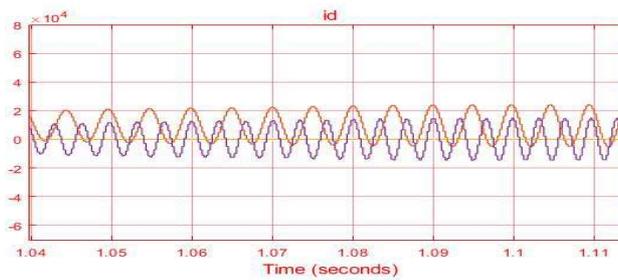


Figure 19

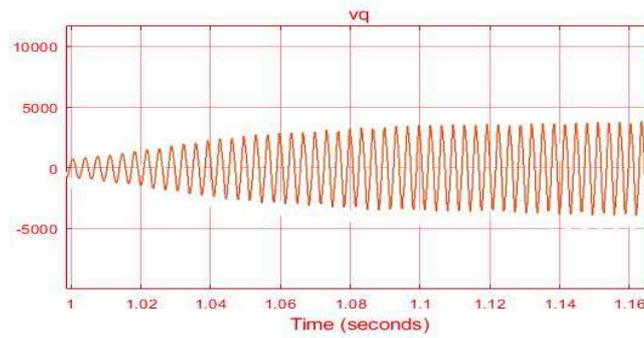


Figure 20

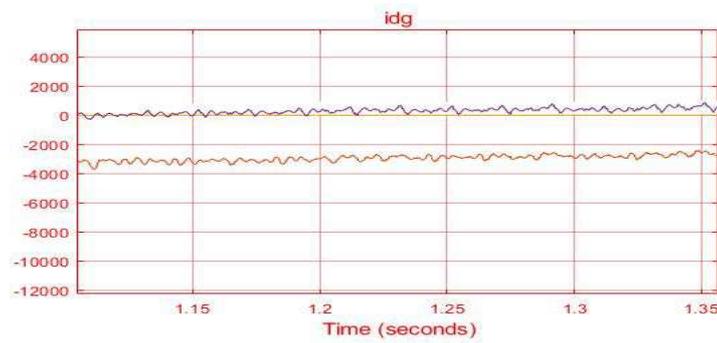


Figure 21



Figure 22

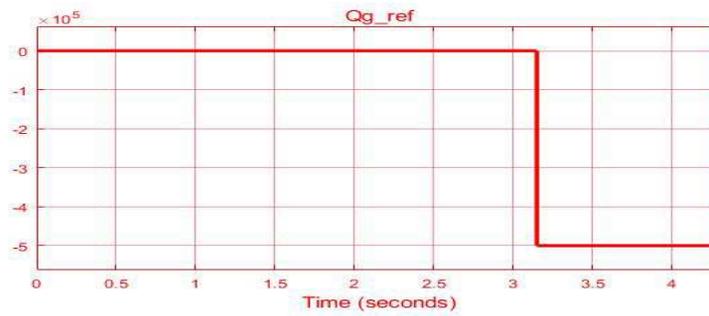


Figure 23

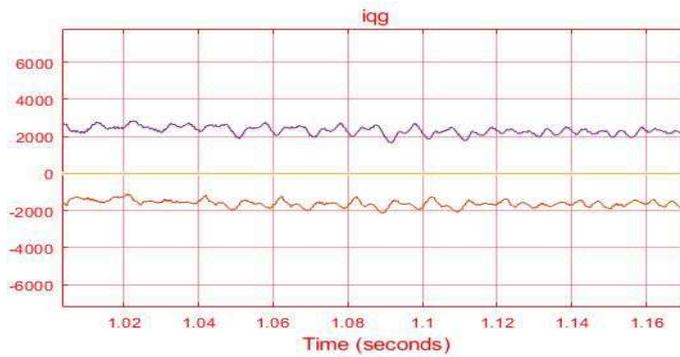


Figure 24

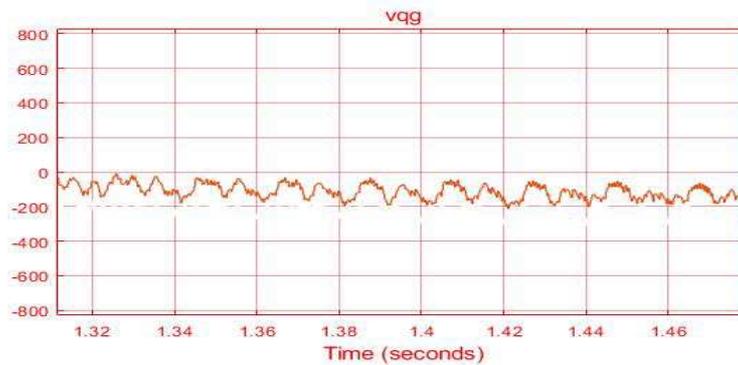


Figure 25

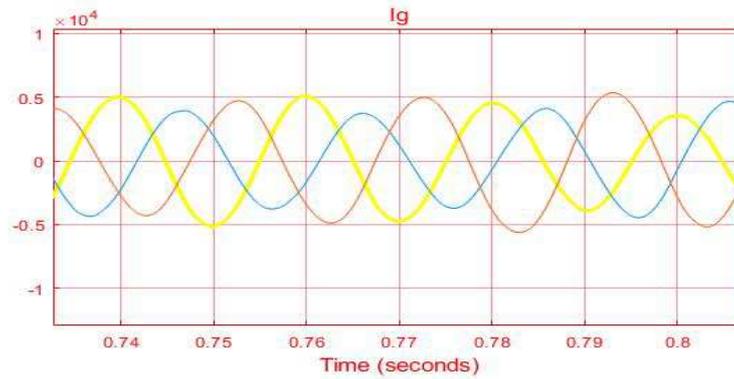


Figure 26

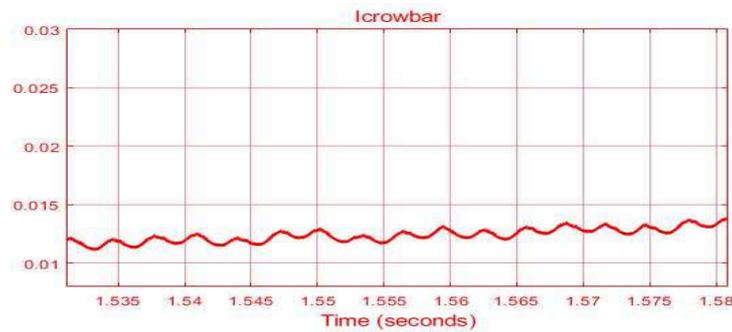


Figure 27

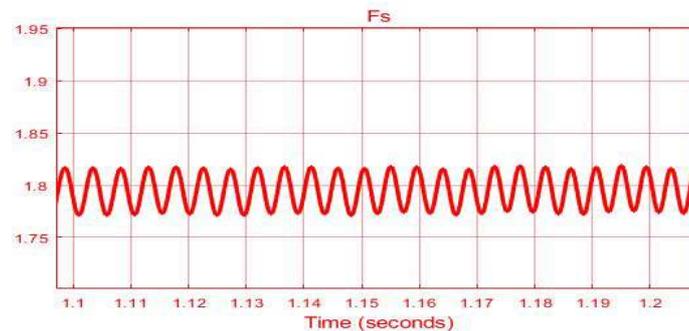


Figure 28

CONCLUSIONS

The DFIG with variable speed ability has higher energy capture efficiency and improved power quality and voltage. The DFIG are more advantages in wind power systems. DFIG wind turbine equipped with Field-oriented control described in this paper is able to provide LVRT and wind turbine stay connected to the grid and limits the rotor currents in an acceptable range. The FOC regulates the active and reactive power by setting reference rotor dq current components in synchronous reference frame. The reference frame can be oriented with stator flux or with grid voltage. Generally, voltage dips are performed at source side. In normal operation stator voltage is reduced to small value, in order to keep this voltage a stator flux can develop a small value. To avoid this problem a crowbar protection is used. In this paper a FOC method topology is explored. The dynamic performance with proposed controller is investigated using voltage sag and step response to reactive power injection. Simulation and experimental results verify the effectiveness and viability of the proposed method.

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